

Testbeam results of irradiated SiGe BiCMOS monolithic silicon pixel detector without internal gain layer

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ABSTRACT: Samples of the monolithic silicon pixel ASIC prototype produced in 2022 within the framework of the Horizon 2020 MONOLITH ERC Advanced project were irradiated with 70 MeV protons up to a fluence of 1×10^{16} n_{eq}/cm², and then tested using a beam of 120 GeV/c pions. The ASIC contains a matrix of 100 μm pitch hexagonal pixels, read out by low noise and very fast frontend electronics produced in a 130 nm SiGe BiCMOS technology process. The dependence on the proton fluence of the efficiency and the time resolution of this prototype was measured with the frontend electronics operated at a power density between 0.13 and 0.9 W/cm². The testbeam data show that the detection efficiency of 99.96% measured at sensor bias voltage of 200 V before irradiation becomes 96.2% after a fluence of 1×10^{16} n_{eq}/cm². An increase of the sensor bias voltage to 300 V provides an efficiency to 99.7% at that proton fluence. The timing resolution of 20 ps measured before irradiation rises for a proton fluence of 1×10^{16} n_{eq}/cm² to 53 and 45 ps at HV = 200 and 300 V, respectively.

KEYWORDS: Particle tracking detectors (Solid-state detectors); Radiation-hard detectors; Solid state detectors

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1 Introduction

To cope with the large event pile up foreseen during the CERN LHC High-Luminosity data-taking period, the experiments showed evidence for the need to upgrade the present detectors with layers with timing capability of the order of tens of picoseconds. The major LHC Collaborations foresee the addition of timing layers [1, 2] to match the required performance. One choice is to build these timing layers with mm² silicon pads based on the low-gain avalanche detectors (LGAD) [3], which feature an internal gain layer under the pixel.

The particle-physics community is presently attempting to develop a new generation of silicon sensors able to achieve both high spatial granularity and excellent timing capabilities [4, 5], although the radiation tolerance of the gain layer still constitutes a problem to place these sensors at small radii in present and future high-energy hadron colliders, where radiation will be very high. Studies to overcome the limited radiation tolerance of the gain layer are ongoing [6–8]. A particularly interesting approach is to use a resistive layer that spreads the signal among four adjacent pixels and, by reconstructing the hit position from those signals, reduces drastically the number of detection channels needed to have spatial resolutions at the level of 10 μm, at the price of a reduction of the timing performance by approximately a factor of two [9–11].

The MONOLITH Horizon 2020 ERC Advanced project utilises the SG13G2 130 nm SiGe BiCMOS process by IHP to produce low noise, low power and very fast frontend electronics, implemented in a fully sensitive high granularity monolithic sensor able to provide excellent timing. The foundry masks of the first prototype of the MONOLITH project [12] were used to produce a proof-of-concept picosecond avalanche detector (PicoAD) [13], a novel detector that implements a continuous deep gain layer [14]. At a power density of 2.7 W/cm², this proof-of-concept monolithic ASIC provided full efficiency and an average time resolution of 17 ps, varying between 13 ps at the center of the pixel and 25 ps in the inter-pixel region [15].

A second prototype of the MONOLITH project containing an improved electronics [16] was produced in 2022. As for the first prototype, the new device contains four pixels where the amplifier was connected to an analog driver and could be read directly by an oscilloscope. Figure 1 shows the layout of the 2022 prototype ASIC with the analog pixels highlighted in red. Figure 2 shows a schematic view of the front-end electronics and analog driver.

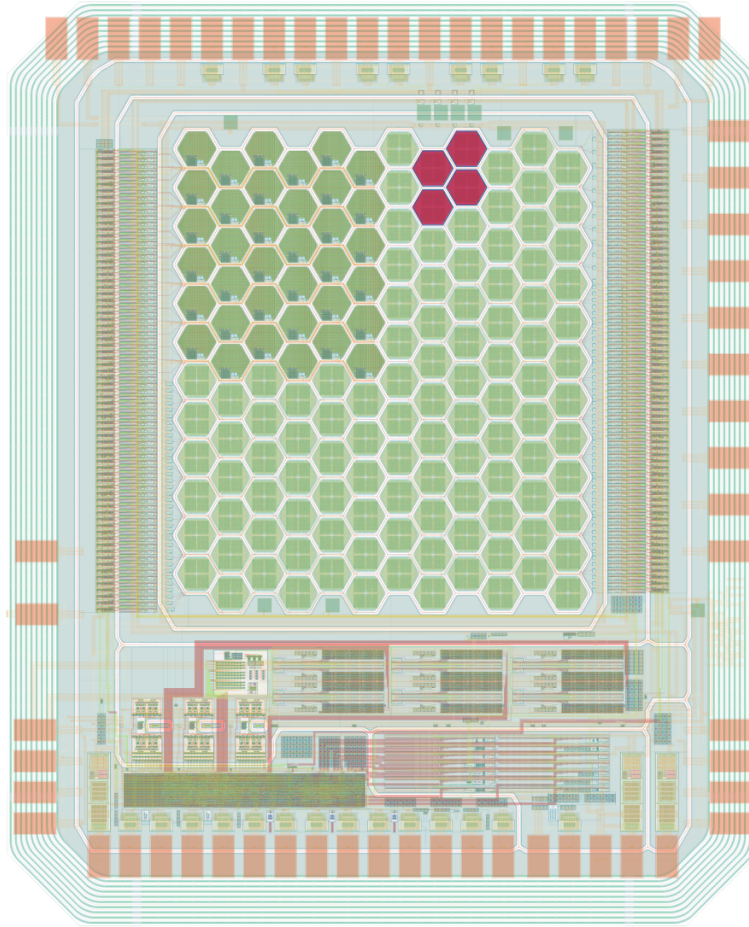


Figure 1. Layout view of the 2022 prototype ASIC. The analog pixels, in red, have the output of the amplifier directly connected to an analog driver with differential output.

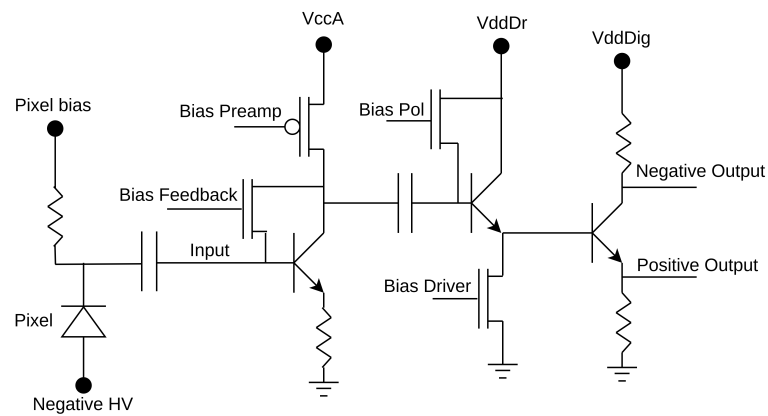


Figure 2. Schematic view of the front-end and driver configuration of the analog pixels.

Before the production of the next generation PicoAD prototypes using special wafers with deep gain-layer implant, the foundry masks of this second prototype were used to realise a detector without internal gain layer, implementing the sensor on a 50 μm thick epilayer of 350 Ωcm resistivity. The

detection efficiency and time resolution of this prototype was measured [16] at the CERN SPS testbeam facility with minimum ionising particles, and the time resolution as a function of the deposited charge was measured [17] with a femtosecond laser. Several samples of this ASIC were irradiated at the CYRIC facility [18] in Japan with 70 MeV protons up to a fluence of $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$. The time jitter of the single-ended output of the analog pixels was measured using a ^{90}Sr radioactive source [19]. In this paper, we present the efficiency and time resolution before and after proton irradiation measured with this prototype without gain layer using a beam of pions of 120 GeV/c at the CERN SPS.

2 Data samples and experimental set-up

Given the limited time availability during the testbeam experiment, data were taken only with four irradiated boards out of the seven characterised in [19]. In addition, data were taken also with the board not irradiated previously characterised in [16]. Table 1 gives a summary of the power density of the frontend and of the sensor bias voltage of the 18 datasets taken at the testbeam. All boards were operated at a frontend feedback current of $2.0 \mu\text{A}$ and, to avoid any unwanted annealing, stored and operated at a temperature of -10°C .

Table 1. Power density and sensor bias voltage of the 18 data samples taken at the CERN SPS testbeam with the four irradiated boards and with the board not irradiated. The proton-fluence values are reported in the first column in $1 \text{ MeV n}_{\text{eq}}/\text{cm}^2$ values.

Proton Fluence [$1 \text{ MeV n}_{\text{eq}}/\text{cm}^2$]	P_{density} [W/cm^2]	High Voltage [V]
0	0.13 0.54 0.9	200
9×10^{13}	0.13 0.54 0.9	
6×10^{14}	0.13 0.54 0.9	
3×10^{15}	0.13 0.54 0.9	
1×10^{16}	0.13 0.54	250
	0.9	150 200 250 300

The UNIGE FE-I4 telescope [20] was used to provide external tracking. The device under test (DUT) was mounted at the center of the telescope as explained in [16]. Two microchannel plate detectors (MCP) were positioned downstream of the telescope and used as time reference. The analog outputs of the two MCP detectors and of the DUT were read by two fast oscilloscopes. The first one with a sampling rate of 20 GS/s and an analog bandwidth set to 4 GHz for the channels connected to the MCPs and 1 GHz for the channels connected to the pair of single-ended analog output of the pixel used to measure the DUT time resolution. Two other pairs of single-ended output of adjacent pixels were connected to the second oscilloscope with a sampling rate of 10 GS/s and an analog bandwidth set to 1 GHz. The DUT was mounted on a cooling plate and covered by an insulating box to keep the ASIC at a constant temperature of -10°C .

The waveforms acquired for the positive and negative single-ended signals of the DUT were subtracted offline, sample point by sample point, to construct differential signals. Throughout the analysis, only the differential signals were used since they provide better signal-to-noise ratio.

3 Detection efficiency

The detection efficiency was computed using pion tracks reconstructed by the FE-I4 telescope that produced hits in the six telescope planes and with a $\chi^2/\text{NDF} \leq 1.5$. Since the number of channels available in the two oscilloscopes allowed recording of the signals from only two out of six neighboring analog pixels, the calculation of the efficiency was restricted to the telescope tracks whose extrapolation on the pixel surface was inside the region defined by the triangle connecting the center of the three pixels (the pixel under study and the two adjacent pixels that were read out). This triangle represents in the correct proportions all the areas of a pixel, and provide a result unbiased by the pointing resolution of the telescope which have been observed to show artificial inefficiency at the edge of pixels if the adjacent pixel is not read out [16].

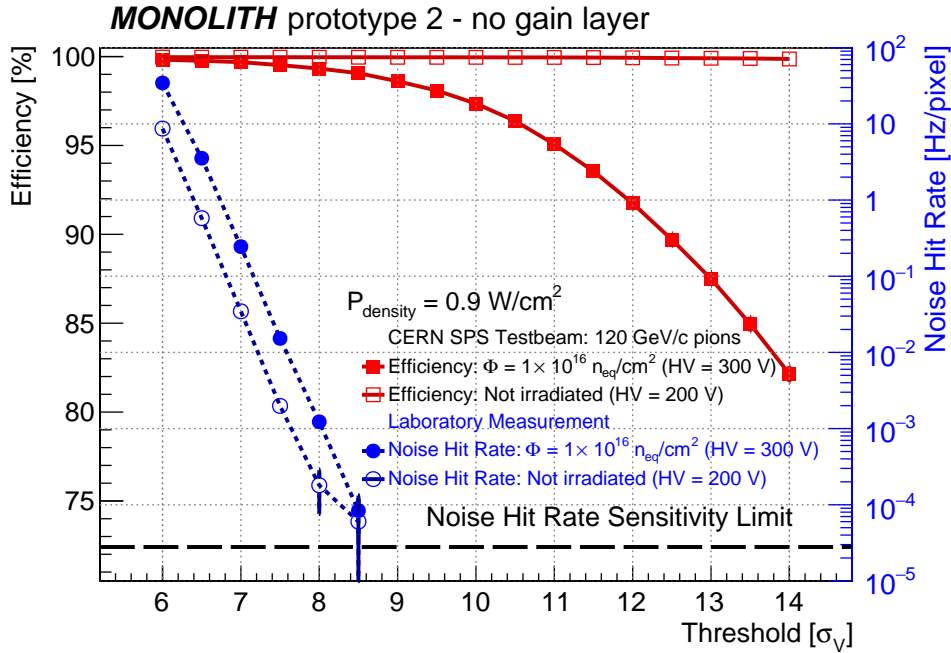


Figure 3. Detection efficiency measured at a power density $P_{\text{density}} = 0.9 \text{ W/cm}^2$ as a function of the voltage threshold. The data refer to the board not irradiated operated at $\text{HV} = 200 \text{ V}$ (open red squares), and to the board irradiated at $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$ operated at $\text{HV} = 300 \text{ V}$ (full red squares). The voltage threshold is given in units of the voltage noise σ_V , which is $300 \mu\text{V}$ before irradiation and raises to $600 \mu\text{V}$ after a fluence of $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$, as reported in [19]. The plot reports also the noise-hit rate (values reported on the right-hand-axis scale) measured with the two boards.

Figure 3 shows the efficiency as a function of the voltage noise σ_V , in the case of the board not irradiated operated at a sensor bias voltage of 200 V , and of the board irradiated at $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$ operated at 300 V . The increase of HV for the irradiated board was necessary to fully deplete the sensor and operate it at very high efficiency after the change in resistivity due to the exposure to radiation. The voltage noise σ_V was measured event-by-event using the waveform samplings recorded in a time interval of 200 ns preceding the signal. At the nominal threshold value $V_{\text{th}} = 7\sigma_V$, the ASIC not irradiated provides a detection efficiency of $(99.96^{+0.01}_{-0.02})\%$ and, given the large signal-to-noise ratio, it maintains an efficiency at the level of 99.8% even with a threshold value of $14\sigma_V$. On the contrary, in

the case of the ASIC that was exposed to $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$, since the signal-to-noise ratio diminishes, the efficiency was found to depend more on the threshold, and at $\text{HV} = 300 \text{ V}$ varies from 99.7% at a threshold $V_{\text{th}} = 7\sigma_V$ to approximately 97.0% already at $V_{\text{th}} = 10\sigma_V$. Figure 3 also reports the noise-hit rate as a function of the threshold. The noise-hit rate shows the expected exponential drop. It increases by a factor of approximately five after the ASIC has received a proton fluence of $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$.

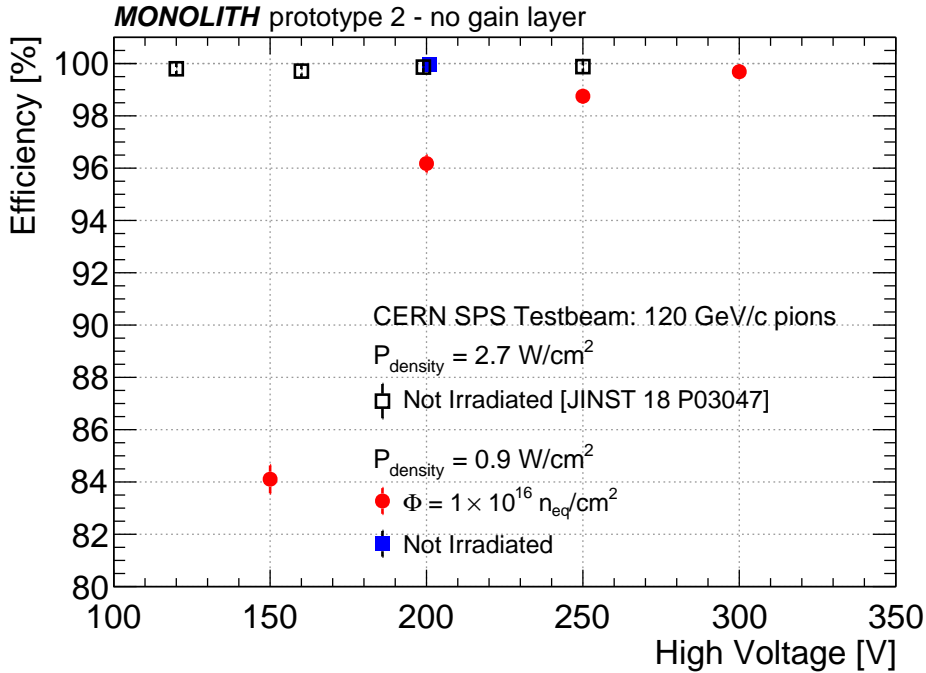


Figure 4. Detection efficiency measured as a function of the sensor bias voltage at a threshold value $V_{\text{th}} = 7\sigma_V$. The red dots show the results obtained with the board irradiated at $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$, while the blue square that obtained with the board not irradiated, all operated at $P_{\text{density}} = 0.9 \text{ W/cm}^2$. The results obtained with the board not irradiated at $P_{\text{density}} = 2.7 \text{ W/cm}^2$ reported in [16] are superimposed as black open squares.

Table 2 reports the detection efficiencies obtained at a threshold $V_{\text{th}} = 7\sigma_V$ for the 18 datasets listed in table 1.

Figure 4 shows the detection efficiency of the ASIC irradiated at $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$ as a function of the sensor bias voltage. Only one data set at $\text{HV} = 200 \text{ V}$ was acquired at the testbeam with the board not irradiated. The data taken with the board not irradiated in a previous testbeam [16] are superimposed in figure 4 as open black squares for comparison.¹ The measurement shows that after $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$ a bias voltage of 300 V is just enough to reach the detection efficiency plateau. This behaviour is completely different from that of the unirradiated board, for which the efficiency plateau is reached already at $\text{HV} = 120 \text{ V}$.

This measurement, as well as the drop in detection efficiency shown in figure 3, could in large part be attributed to the change in resistivity of the silicon bulk, that we estimated to vary from the initial $350 \text{ }\Omega\text{cm}$ to approximately $50 \text{ }\Omega\text{cm}$ extrapolating the data of [21, 22]. At such resistivity, the $50 \text{ }\mu\text{m}$

¹It should be noted that, in addition to the different power density, the data from [16] were acquired at feedback current $i_{\text{feedback}} = 0.1 \text{ }\mu\text{A}$ and at a temperature of 20°C in contrast with the $2.0 \text{ }\mu\text{A}$ and -10°C of the present data. This change in operating conditions results in an increase of the signal-to-noise ratio, that leads to better efficiency and time resolution.

Table 2. Detection efficiency and time resolution measured for the 18 data samples taken at the CERN SPS testbeam with the four irradiated boards and with the board not irradiated. The values were obtained for a signal amplitude threshold $V_{th} > 7\sigma_V$.

Proton Fluence [1 MeV n_{eq}/cm^2]	$P_{density}$ [W/cm ²]	High Voltage [V]	Detection Efficiency [%]	Time Resolution [ps]
0	0.13	200	$99.94^{+0.03}_{-0.05}$	31.0 ± 1.6
	0.54		$99.96^{+0.02}_{-0.03}$	29.0 ± 1.3
	0.9		$99.96^{+0.01}_{-0.02}$	20.0 ± 1.0
9×10^{13}	0.13	200	$99.12^{+0.10}_{-0.19}$	37.8 ± 2.8
	0.54		$99.63^{+0.07}_{-0.13}$	29.0 ± 1.3
	0.9		$99.78^{+0.04}_{-0.11}$	24.7 ± 1.0
6×10^{14}	0.13	200	$98.38^{+0.17}_{-0.19}$	42.4 ± 3.0
	0.54		$99.13^{+0.12}_{-0.13}$	29.9 ± 2.2
	0.9		$99.24^{+0.10}_{-0.11}$	26.4 ± 1.5
3×10^{15}	0.13	200	$94.06^{+0.22}_{-0.23}$	75.9 ± 4.5
	0.54		$97.22^{+0.17}_{-0.18}$	59.4 ± 0.9
	0.9		$97.97^{+0.11}_{-0.11}$	48.3 ± 1.9
1×10^{16}	0.13	250	$94.74^{+0.18}_{-0.19}$	74.0 ± 5.4
	0.54		$98.33^{+0.10}_{-0.11}$	48.3 ± 1.9
	0.9	150	$84.11^{+0.53}_{-0.54}$	60.4 ± 3.3
		200	$96.18^{+0.31}_{-0.33}$	53.1 ± 3.4
		250	$98.75^{+0.09}_{-0.10}$	50.2 ± 1.6
		300	$99.69^{+0.06}_{-0.07}$	45.3 ± 1.6

thick sensor would not be fully depleted. A concurrent radiation damage of the frontend electronics cannot be excluded, and should be measured irradiating the SiGe HBT alone.

Figure 5 shows the impact on the detection efficiency of the power density at which the frontend was operated. At a sensor bias voltage $HV = 200$ V, a steady increase of the efficiency is measured as $P_{density}$ increases. This effect is more pronounced at the highest proton fluences.

In the case of the board irradiated to 1×10^{16} n_{eq}/cm^2 , at $HV = 200$ V data were taken only at $P_{density} = 0.9$ W/cm², which result in a detection efficiency of $(96.2 \pm 0.3)\%$ (red circle in the figure). For this board the scan in $P_{density}$ was performed at $HV = 250$ V, and the results are displayed in figure 5 by the red crosses. At $HV = 300$ V, only a dataset at $P_{density} = 0.9$ W/cm² was acquired (red star in the figure), which provides a detection efficiency of $(99.69^{+0.06}_{-0.07})\%$.

Finally, the detection efficiency is shown as a function of the distance from the centre of the pixel in figure 6. All the data points shown refer to a sensor bias voltage of 200 V, with the exception

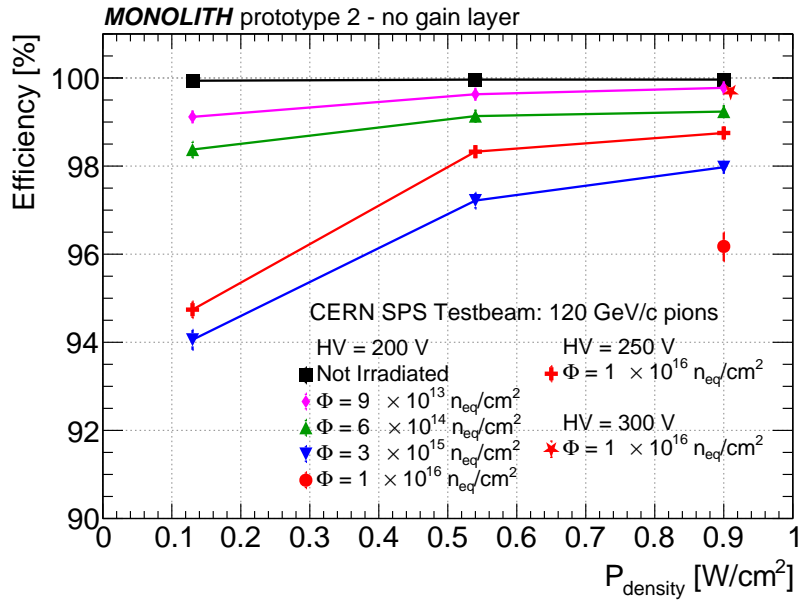


Figure 5. Detection efficiency measured for power density supplied to the preamplifier between 0.13 and 0.9 W/cm² for a threshold value of $V_{th} = 7\sigma_V$. The results obtained with the sample not irradiated (black empty squares) are presented with those of the four samples irradiated with fluences between 9×10^{13} n_{eq}/cm² and 1×10^{16} n_{eq}/cm², all operated with a sensor bias voltage of HV = 200 V. In the case of the board irradiated at 1×10^{16} n_{eq}/cm² the data taken at sensor bias voltage of 250 and 300 V are also shown. The marker relative to the data point at HV = 300 V (red star) is displayed with a small horizontal offset to avoid overlapping with other markers.

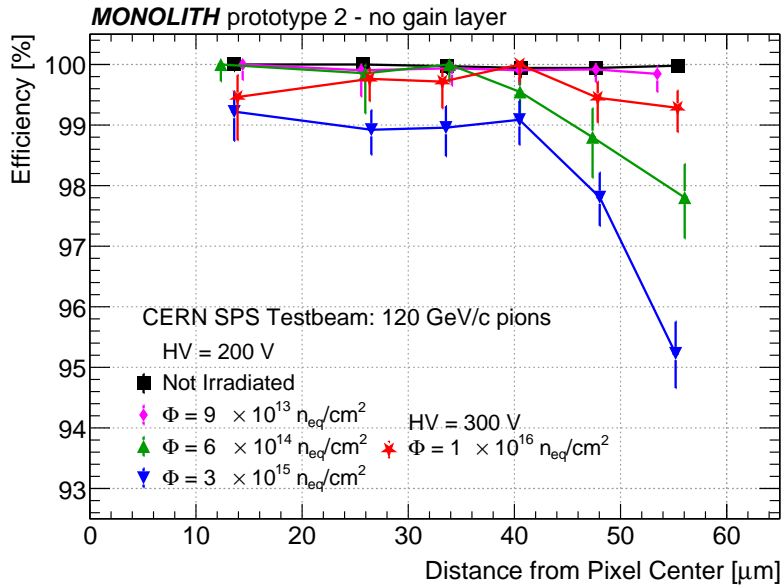


Figure 6. Detection efficiency as a function of the distance from the pixel center. The data refer to power density $P_{density} = 0.9$ W/cm², voltage threshold $V_{th} = 7\sigma_V$ and sensor bias voltage HV = 200 V, with the exception of the data taken at 1×10^{16} n_{eq}/cm², for which the sensor bias voltage was 300 V. The efficiency was computed in the following bins of distance from the pixel center in microns: [0–21], [21–30], [30–37], [37–44], [44–51], [51–65]. In each bin, the data points are plotted at the mean value of the track distance from the pixel center.

of the dataset at $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$, which was obtained at 300 V. While at no or small irradiation values the efficiency does not depend on the position within the pixel area, starting from a fluence of $6 \times 10^{14} \text{ n}_{\text{eq}}/\text{cm}^2$ the efficiency was measured to drop in the inter-pixel area, although an increase of the sensor bias voltage allows recovering of the efficiency far from the pixel center, as the data at $\text{HV} = 300 \text{ V}$ (red stars) show.

4 Time resolution

The time resolution of the DUT was measured using as reference the timestamp of two precise MCP detectors that were located outside the FE-I4 telescope downstream the pion beam. The sample of telescope tracks used for the measurement of the time resolution was the same used for the detection efficiency, with the only exception that all the tracks within the pixel were considered here, and not exclusively the tracks within the aforementioned triangle defined by the three analog pixel centers.

The time of arrival (TOA) was measured for the DUT and the MCPs making offline a linear interpolation between the oscilloscope samplings and taking the time at which the signals reached 50% of the maximum value of the signal amplitude. The difference in TOA between the three pairs of detectors was then used to extract the time resolution of the DUT, as explained in [16].

Taking the TOA at a fixed percentage of the signal amplitude has the advantage to correct for most of the time walk between signals with different amplitudes, under the assumption that signals of all amplitudes have consistent rise times. It was observed that in our case this method carries the drawback of creating a small distortion of the distribution of the TOA difference, which deviates slightly from a Gaussian distribution and generates a tail at large values of $\text{TOA}_{\text{DUT}} - \text{TOA}_{\text{MCP}}$, probably from residual time walk not corrected for. This effect is shown in figure 7 in the case of the board not irradiated (left panels) and of the board irradiated at $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$ (right panels).

To account for the observed distortion in the TOA difference, the σ values from the Gaussian fits obtained using only the bins of the distribution larger than 20% of the bin with the highest number of entries (two top panels in the figure 7) were used as central values of the time resolution. The difference between these values and those obtained by fitting a Gaussian functional form in the entire TOA-difference distributions (bottom panels in figure 7) were used as systematic uncertainty. These uncertainties were summed in quadrature with the statistical uncertainty coming from the Gaussian fits that were used as central values, and are shown as vertical error bars in the time resolution plots.

The last column of table 2 reports the time resolution obtained for the 18 datasets acquired at $V_{\text{th}} = 7\sigma_V$.

Figure 8 shows the time resolution obtained at $P_{\text{density}} = 0.9 \text{ W}/\text{cm}^2$ for the board irradiated to $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$ for a sensor bias voltage varying between 150 and 300 V, and the result obtained with the unirradiated board at $\text{HV} = 200 \text{ V}$. For comparison, the figure reports as well the time resolution measured at $P_{\text{density}} = 2.7 \text{ W}/\text{cm}^2$ with the board not irradiated in previous testbeam experiment [16]. While the board not irradiated reaches the plateau of time resolution already at $\text{HV} = 160 \text{ V}$, the time resolution measured with the irradiated board does not seem to have yet reached a plateau value at 300 V.

The time resolution of the DUT was also measured as a function of the power density supplied to the frontend electronics. The results are presented in figure 9 at a sensor bias voltage $\text{HV} = 200 \text{ V}$ for all the boards, and for the most irradiated board also at $\text{HV} = 250$ and 300 V . In all cases, the time resolution was found to steadily improve with increasing P_{density} , and degrade with increasing

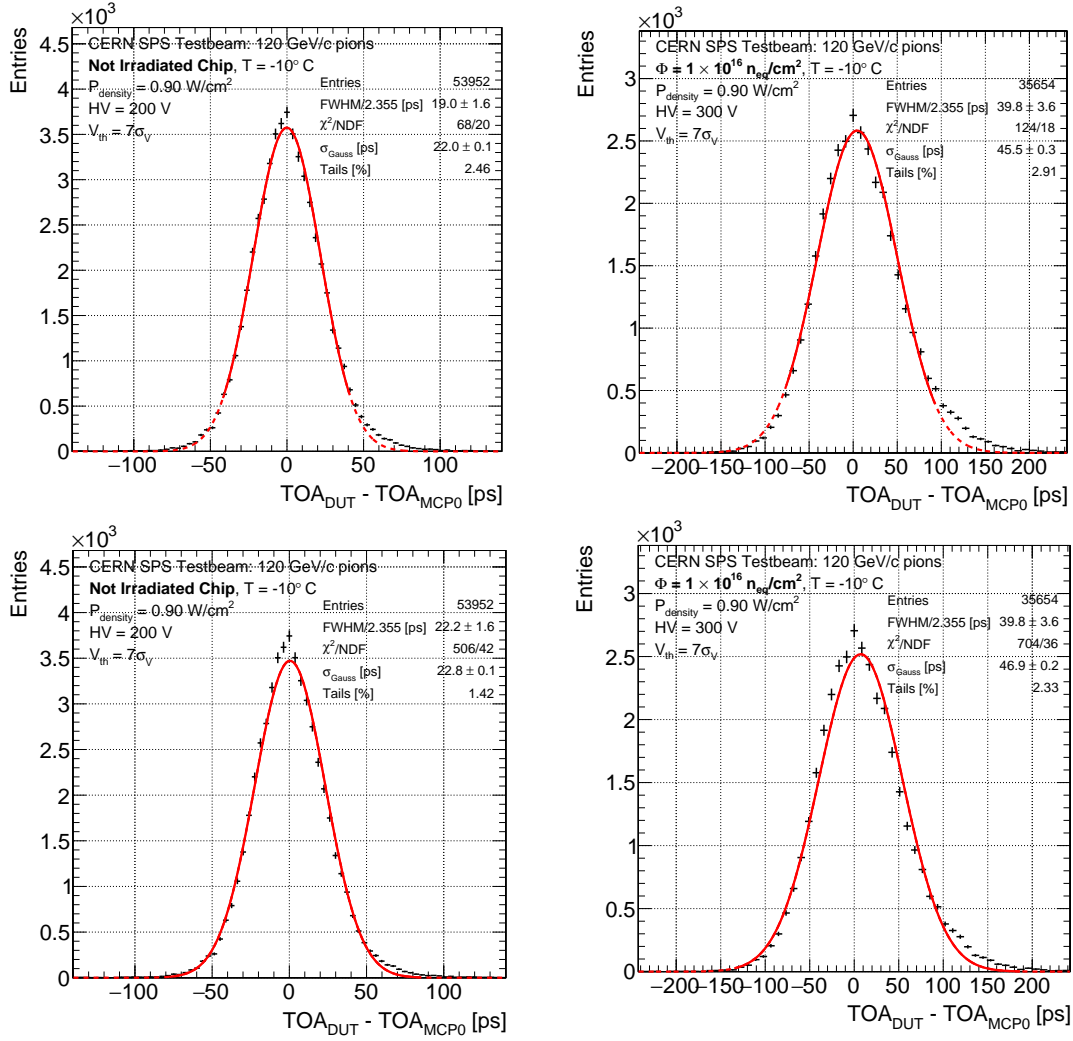


Figure 7. Difference in TOA between the DUT and one of the MCP detectors used to provide a precise reference time. The left-hand panels show the results for the board not irradiated, while the right-hand panels those for the board irradiated to $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$. The top panels show the data with superimposed the Gaussian fits extending to the first bin containing 20% of the maximum value of the distribution; the bottom panels show the same data with superimposed the Gaussian fits extending to all the bins of the distributions.

proton fluence. In the P_{density} range studied, it varies between 30 and 20 ps in the case of the board not irradiated, and between 75 and 50 ps in the case of the board irradiated to $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$.

It should be noted that, for each of the irradiation values studied, the ratio of the time resolutions at $P_{\text{density}} = 0.13$ and $0.9 \text{ W}/\text{cm}^2$ is constant, showing a relative improvement of approximately 35% independently of the fluence, as reported in table 3.

This is an indication that the dependence of the transition frequency from the collector current, which is observed in [23] not to change up to $1 \times 10^{14} \text{ n}_{\text{eq}}/\text{cm}^2$, might not be affected by radiation damage even at $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$.

Figure 10 shows the time resolutions as a function of the proton fluence at $P_{\text{density}} = 0.9 \text{ W}/\text{cm}^2$. The results obtained with the datasets taken at sensor bias voltage HV = 200 V vary between 20 ps for the board not irradiated and 53 ps for the board irradiated to $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$. For this last

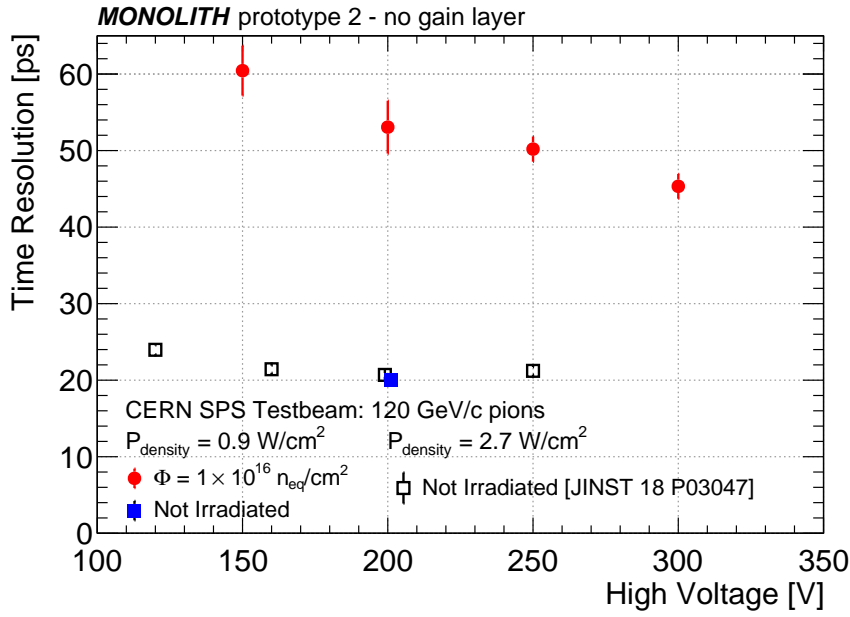


Figure 8. Time resolution measured as a function of the sensor bias voltage. The results obtained with the board irradiated to $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$ at $P_{\text{density}} = 0.9 \text{ W/cm}^2$ (red circles) are presented together with the result obtained with the board not irradiated at $\text{HV} = 200 \text{ V}$ (blue square). The plot results obtained in a previous testbeam [16] with the board not irradiated at $P_{\text{density}} = 2.7 \text{ W/cm}^2$ and $i_{\text{feedback}} = 0.1 \mu\text{A}$ are also superimposed (black open squares).

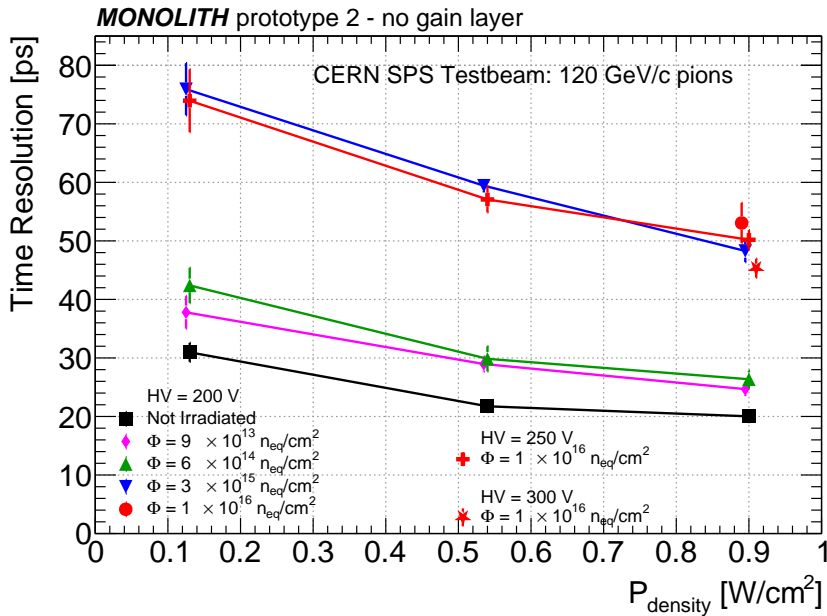


Figure 9. Time resolution measured for three values of the power density supplied to the preamplifier between 0.13 and 0.9 W/cm^2 . The plot shows the results obtained at sensor bias voltage $\text{HV} = 200 \text{ V}$ with the five boards which were subject to proton fluence between zero and $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$. The data taken with the board most irradiated at $\text{HV} = 250$ and 300 V are also shown.

Table 3. Time resolution measured with the five boards at $P_{\text{density}} = 0.13$ and 0.9 W/cm^2 and their ratio. The data were taken at sensor bias voltage $\text{HV} = 200 \text{ V}$ for all boards except the board irradiated to $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$ for which the sensor bias voltage was 250 V .

Fluence [$1 \text{ MeV n}_{\text{eq}}/\text{cm}^2$]	Time Resolution [ps]		Ratio
	$P_{\text{density}} = 0.13 \text{ W/cm}^2$	$P_{\text{density}} = 0.9 \text{ W/cm}^2$	
0	31.0 ± 1.6	20.0 ± 1.0	1.55 ± 0.15
9×10^{13}	37.8 ± 2.8	24.7 ± 1.0	1.53 ± 0.21
6×10^{14}	42.4 ± 3.0	26.4 ± 1.5	1.61 ± 0.21
3×10^{15}	75.9 ± 4.5	48.3 ± 1.9	1.57 ± 0.17
1×10^{16}	74.0 ± 5.4	50.2 ± 1.6	1.57 ± 0.19

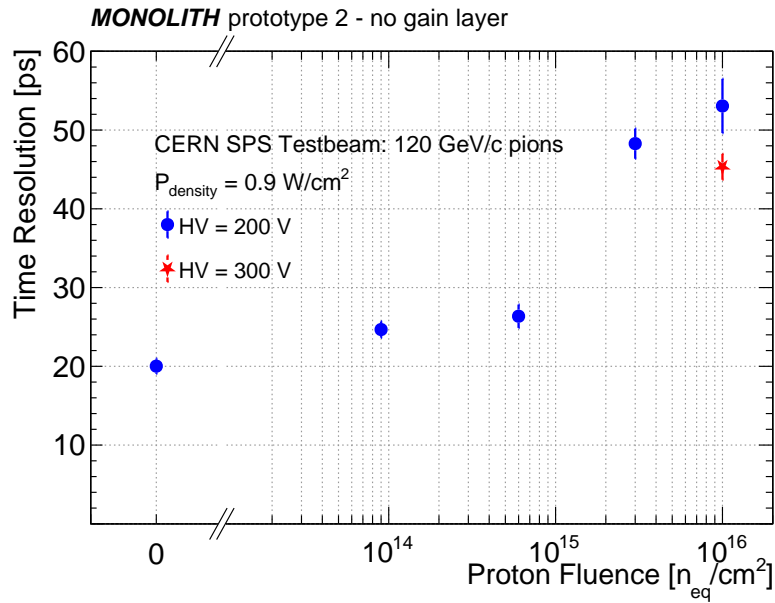


Figure 10. Time resolution as a function of the proton fluence for $P_{\text{density}} = 0.9 \text{ W/cm}^2$. The blue dots show the results obtained at a sensor bias voltage $\text{HV} = 200 \text{ V}$, while the red star shows the result obtained at $\text{HV} = 300 \text{ V}$ with the board irradiated to $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$.

board, an increase of the sensor bias voltage to 300 V provides a time resolution of 45 ps , thus improving it by approximately 10% .

The time resolution is shown as a function of the distance from the centre of the pixel in figure 11. All the data refer to a sensor bias voltage of 200 V , with the exception of the dataset at $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$, which was acquired at 300 V . For all fluences the time resolution is better in the central region of the pixel than in the inter-pixel region, where it increases approximately 5 ps .

5 Summary and conclusions

A monolithic silicon pixel prototype produced for the Horizon 2020 MONOLITH ERC Advanced project was irradiated up to a proton fluence of $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$. The prototype contains a matrix of

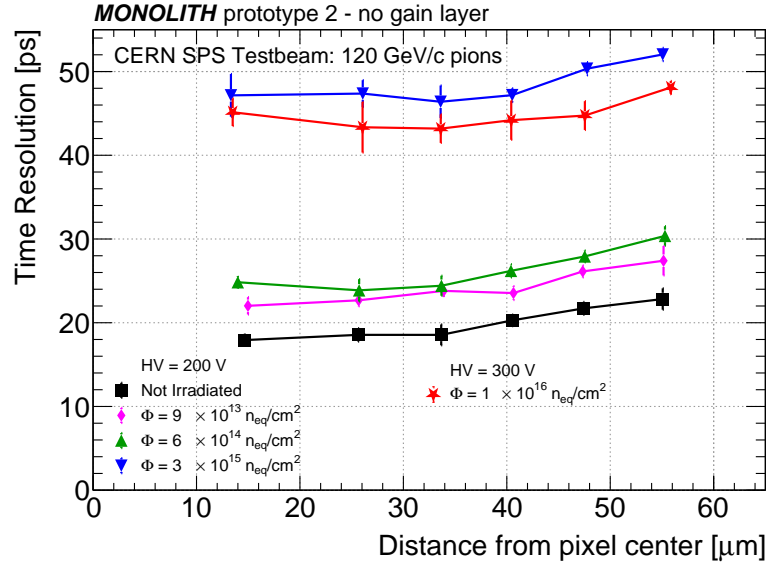


Figure 11. Time resolution as a function of the distance from the pixel center. The data refer to $P_{\text{density}} = 0.9 \text{ W/cm}^2$, $V_{\text{th}} = 7\sigma_V$ and $\text{HV} = 200 \text{ V}$, with the exception of the data taken at $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$, for which the sensor bias voltage was 300 V . The time resolution was computed in the same bins of figure 6.

hexagonal pixels with $100 \mu\text{m}$ pitch, read by a SiGe BiCMOS 130 nm frontend electronics. The sensor was produced on a $50 \mu\text{m}$ epilayer of a resistivity of $350 \Omega\text{cm}$. The analog channels present in the pixel matrix were read by a fast oscilloscope and tested at the CERN SPS using a beam of $120 \text{ GeV}/c$ pions. Only signals with an amplitude larger than a voltage threshold $V_{\text{th}} = 7\sigma_V$ were considered for the measurement of the detection efficiency and time resolution.

Before proton irradiation, the detection efficiency at a sensor bias voltage $\text{HV} = 200 \text{ V}$ and frontend power density $P_{\text{density}} = 0.9 \text{ W/cm}^2$ was measured to be $(99.96_{-0.02}^{+0.01})\%$. For a proton fluence $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$, the detection efficiency in the same conditions was measured to be $(96.2 \pm 0.3)\%$; an increase of the bias voltage to 300 V provided $(99.69_{-0.07}^{+0.06})\%$. These efficiency values include also the inter-pixel regions. Therefore there is no inactive area in between adjacent pixels within the sensor matrix.

The time of arrival of the signals used in the data analysis was defined as the time at 50% of the signal amplitude, and was computed by a linear interpolation of the oscilloscope samplings. No further time-walk correction was applied to the data. The time resolution at $P_{\text{density}} = 0.9 \text{ W/cm}^2$ and $\text{HV} = 200 \text{ V}$ was measured to be $(20.0 \pm 1.0) \text{ ps}$ before proton irradiation. At the highest proton fluence studied of $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$, the time resolution became $(53.1 \pm 3.4) \text{ ps}$. An increase of the sensor bias voltage to $\text{HV} = 300 \text{ V}$ improved the result to $(45.3 \pm 1.6) \text{ ps}$.

The scan in sensor bias voltage of the board exposed to $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$, showed that the plateau in detection efficiency was just reached at 300 V , while the time resolution did not yet reach its plateau value. These observations, supported by an estimation of the change in resistivity after $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$, imply that the sensor was not yet fully depleted at 300 V and suggest that the nominal working point at that proton fluence is beyond 300 V .

In conclusion, this study gives first hints that an amplifier using SiGe HBT is not affected greatly by doses up to $1 \times 10^{16} \text{ n}_{\text{eq}}/\text{cm}^2$. The data show that very high efficiency, very high time resolution

and radiation tolerance can be obtained by monolithic pixelated silicon detectors without internal gain layer that exploit low noise, low power and fast response SiGe HBTs. Commercial SiGe BiCMOS foundry processes therefore represent a very strong candidate to build monolithic silicon detectors for 4D tracking systems at future colliders at a contained cost.

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